



جامعة بنها - كلية الهندسة ببها - قسم الهندسة الكهربائية  
 نموذج الإجابة امتحان مادة هندسة التحكم ك1236 تخلفات مايو يوم الاربعاء الموافق 2016-5-25  
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Benha University	Time: 3-hours	 
Benha Faculty of Engineering	Second Year 25-5-2016	
Control Engineering (E1236)	Elect.Eng.Dept. تخلفات	

**Solve as much as you can questions in two pages**

**Q1**

**(20marks)**

- a-Write a **mathematical model** represents the physical systems shown in Fig.1, and Fig.2?
- b- Draw a block diagram represents the system shown in Fig.2 and using block reduction method to find  $Y(S)/U(S)$ ?
- c-Write the most important features of **a good** control system?
- d-Write the most important **advantages and disadvantages** of the **open loop** and the **closed loop** control systems?

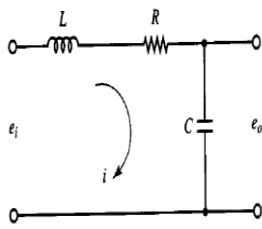


Fig.1

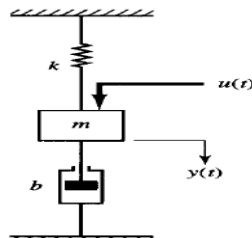


Fig.2

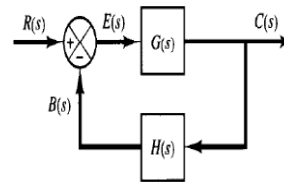


Fig.3

**Q2**

**(20 marks)**

Consider a system shown in Fig. 3  $H(s) = 1$ ,  $G(s) = \frac{K}{s(s+4)}$

- a-Find the steady state static **error coefficients**?
- b- Find the gain **K** such that the steady state error =0.02?
- c- Find and draw the **unit step response** as  $K=16$ ?
- d- Find the **frequency response** and  $M_r$  and  $\omega_r$  as  $K=16$  and  $r(t)=2\sin \omega t$  ?

**P.T.O**

Q3

(20 marks)

Consider a control system shown in Fig.3 if

$$G(s)H(s) = \frac{k}{(S+1+j)(S+2)(S+1-j)} = \frac{k}{S^3+4S^2+6S+4}$$

- Sketch the **complete root locus** for positive values of **K**?
- Find **K** that makes the complex closed loop poles have a damping ratio =**0.5** and **find the closed loop poles using the plot**?
- Find **K** that makes the complex closed loop poles have a damping ratio =**0.5** and **find the closed loop poles analytically**?
- Write short MATLAB program to solve a & b?

Q4

(30 marks)

- Define: gain margin- phase margin?
- Define:  $\omega_n$ ,  $\omega_d$ ,  $\omega_r$ ,  $\omega_c$ ,  $\omega_p$ ,  $M_r$ ,  $\eta$ ?
- Consider a control system shown in Fig.3 if

$$G(s)H(s) = \frac{10}{(S + 1 + j)(S + 2)(S + 1 - j)} = \frac{10}{S^3 + 4S^2 + 6S + 4}$$

- Prove that the gain margin=**6.02 db at 2.45 rad/sec.** and the phase margin=**30.3 degrees at 1.78 rad/sec.**?
- Sketch the **polar plot**?
- Sketch the **Bode plot**?
- Show the gain margin and the phase margin on **the plots**?
- Write short MATLAB program to solve a , b and C?

### Answer

Q1

(20marks)

a-Write a **mathematical model** represents the physical systems shown in Fig.1, and Fig.2?

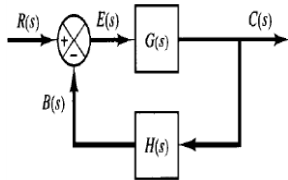
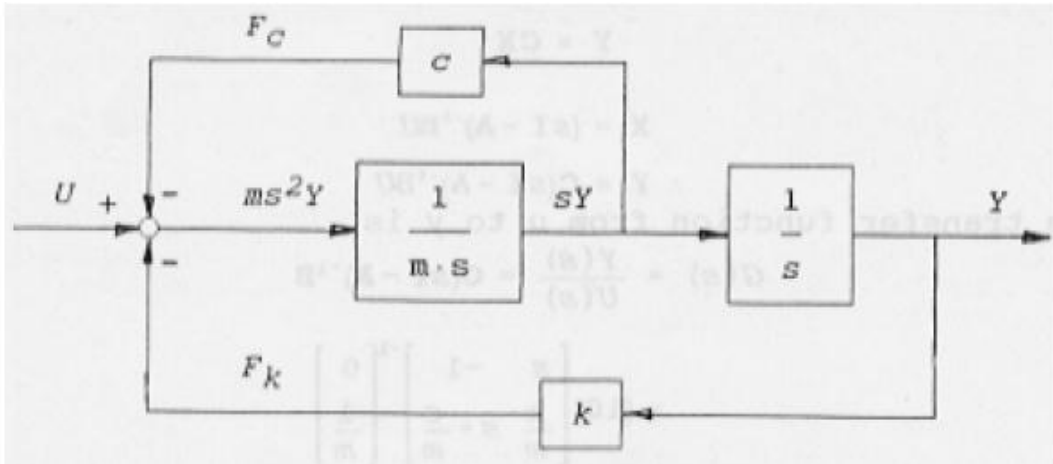
Fig.1

$$\sum_1^n V_{loop} = 0 \text{ then } e_i = R \frac{dq}{dt} + L \frac{d^2q}{dt^2} + \frac{q}{C}, \quad i = \frac{dq}{dt}, \quad e_o = \frac{\int i dt}{C} = \frac{q}{C}$$

Fig.2

$$ma = \sum F = m\ddot{y} = U - b\dot{y} - Ky,$$

b- Draw block diagram represents the system shown in Fig.2 and using block reduction method to find  $Y(S)/U(S)$ ?



$K+SC=H(s), G(s)=1/mS^2$

$$\text{Closed loop tf} = \frac{C(S)}{R(S)} = \frac{G(s)}{1 + G(S)H(S)} = \frac{1}{mS^2 + Sb + K}$$

c-Write the most important features of **a good** control system?

Most important features of a good control system: are

- 1-simple construction and operation
- 2-fast response (speed)
- 3-less cost
- 4-very large accuracy (less error)
- 5-stable

d-Write the most important **advantages and disadvantages** of the **open** loop and the **closed** loop control systems?

**Open loop control system**

Advantages of open loop	disadvantages of open loop
1-simple construction	1-disturbances cause errors
2- ease of maintenance	2-changes in calibration cause errors
3-less expensive	3-recalibration is necessary
4-no stability problem	
5-convenient when output is hard to measured or economically not feasible	

### Closed loop control system

Disadvantages of closed loop	advantages of closed loop
1-complex construction	1-disturbances do not cause errors
2- stability may be a problem	2- has less errors
3-more expensive	3-recalibration is not necessary
	4-the ability to adjust the response

### Q2

(20 marks)

Consider a system shown in Fig. 3 as  $H(s) = 1$ ,  $G(s) = \frac{K}{s(s+4)}$

a- Find the steady state static **error coefficients**?

It must convert non-unity feedback to unity feedback as

$$K_p = \lim_{s \rightarrow 0} G(s) = \lim_{s \rightarrow 0} \frac{K}{s(s+4)} = \frac{K}{(0)(0+4)} = \infty$$

$$K_v = \lim_{s \rightarrow 0} sG(s) = \lim_{s \rightarrow 0} \frac{K}{(s+4)} = \frac{K}{(0+4)} = 0.25K$$

$$K_a = \lim_{s \rightarrow 0} s^2G(s) = \lim_{s \rightarrow 0} \frac{sK}{(s+4)} = 0$$

b-Find the gain **K** such that the steady state error =0.02?

$$e_{ss}(t) = \frac{1}{K_v} = \frac{1}{0.25K} = 0.02, K = 200 \text{ Routh test as } K=200, \text{ system is stable}$$

c-Find and draw the **unit step response** as  $K=16$ ?

$$\frac{C(s)}{R(s)} = \frac{G(s)}{1 + G(s)H(s)} = \frac{\omega_n^2}{s^2 + 2\eta\omega_n s + \omega_n^2} = \frac{16}{s^2 + 4s + 16}, \omega_n = 4 \text{ rad/sec.}, \eta = 0.5$$

The step response is under damped Step response of a second order system  $R(s)=1/s$

$$C(s) = \text{closed loop T.F.}(s) * R(s) = \frac{\omega_n^2 R(s)}{s(s^2 + 2\eta\omega_n s + \omega_n^2)} = \frac{16}{s(s^2 + 4s + 16)}$$

$$= \frac{a}{s} + \frac{bs+d}{(s^2+4s+16)} \text{ partial fraction ,}$$

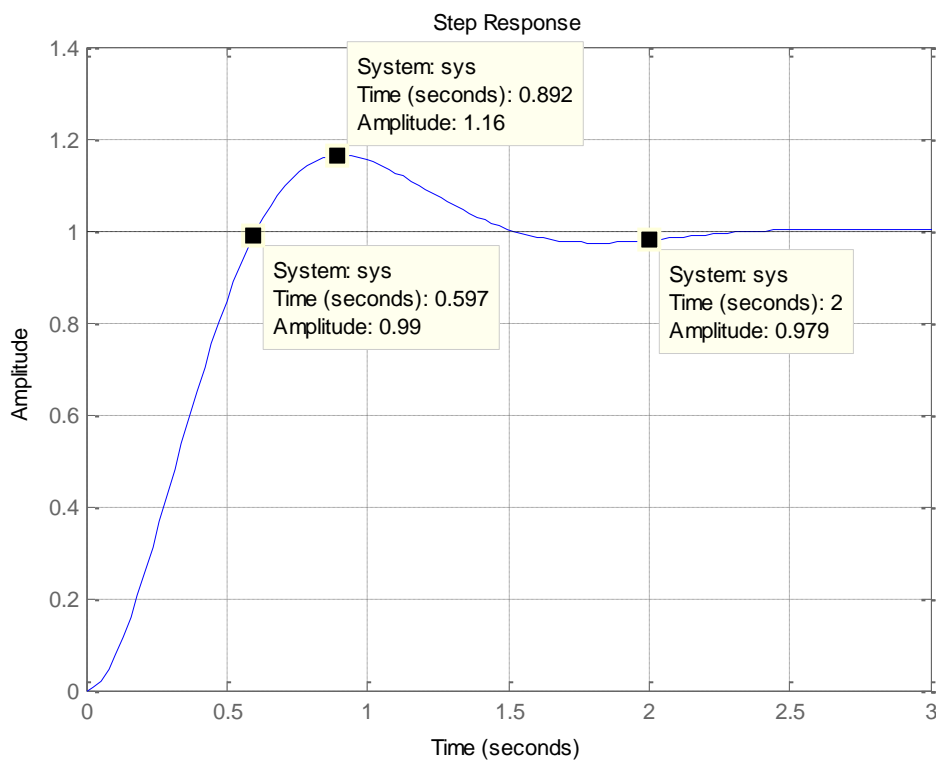
$C(t)$ = inverse Laplace of the product of closed loop t.f.( $S$ ) and  $R(S)=1/s$  with zero initial conditions  $C(t)=L^{-1}[(C(S))]=L^{-1}[\text{closed loop t.f.}(S)*R(S)]$  with zero initial conditions]

$$\eta = 0.5, \omega_n = 4 \text{ rad/sec. } \omega_d = \omega_n \sqrt{1 - \eta^2} = 3.5 \text{ rad/sec } , \cos^{-1} 0.5 = \pi/3$$

$$C(t) = 1 - \frac{e^{-\eta \omega_n t}}{\sqrt{1-\eta^2}} \sin(\omega_d t + \cos^{-1}\eta) = 1 - 1.155e^{-2t} \sin(1.732t + \pi/3)$$

$$M_p = e^{\frac{-\eta\pi}{\sqrt{1-\eta^2}}} = 0.163, t_r = \frac{\pi - \cos^{-1}\eta}{\omega_d} = \frac{\pi - \frac{\pi}{3}}{3.5} = 0.6 \text{ sec},$$

$$t_p = \frac{\pi}{\omega_d} = 0.9 \text{ sec}, t_s = 4T = \frac{4}{\eta\omega_n} = 2 \text{ sec}.$$



Find the **frequency response** and  $M_r$  and  $\omega_r$  as  $K=4$  and  $r(t)=5 \sin \omega t$  ?

d-Find the **frequency response** and  $M_r$  and  $\omega_r$  as  $K=16$  and  $r(t)=2 \sin \omega t$  ?

$$\frac{C(S)}{R(S)} = \frac{G(S)}{1 + G(S)H(S)} = \frac{\omega_n^2}{S^2 + 2\eta\omega_n S + \omega_n^2} = \frac{16}{S^2 + 4S + 16}$$

$$\omega_n = 4 \text{ rad/sec}, \eta = 0.5, M_r = \frac{1}{2\zeta \sqrt{1-\zeta^2}} = \frac{1}{2(0.5) \sqrt{1-(0.5)^2}} = 1.155$$

$$\omega_r = \omega_n \sqrt{1 - 2\zeta^2} = 4\sqrt{1 - 2(0.5)^2} = 2.818 \text{ rad/sec.}$$

the frequency response as  $r(t)=2\sin\omega t$  . Steps to find frequency Response:

1- the closed loop transfer function =  $T(s)=C(S)/R(S) =$

$$\frac{C(S)}{R(S)} = \frac{G(s)}{1 + G(S)H(S)} = \frac{\omega_n^2}{S^2 + 2\eta\omega_n S + \omega_n^2} = \frac{16}{S^2 + 4S + 16}$$

2-the closed loop frequency transfer function =

$$T(j\omega) = \frac{C(j\omega)}{R(j\omega)} = \frac{16}{(j\omega)^2 + 4(j\omega) + 16} = M \angle \Phi = \text{Re} + j \text{imag}$$

$$M = \frac{16}{\sqrt{(16 - \omega^2)^2 + 16\omega^2}}, \quad \Phi = \tan^{-1}[4\omega / (16 - \omega^2)]$$

3-As the input  $=r(t) = 3\sin\omega t$  then

$$\text{the response} = C(t) = 2M\sin(\omega t + \Phi)$$

$$= \frac{32}{\sqrt{(16 - \omega^2)^2 + 16\omega^2}} \sin[\omega t + \tan^{-1}[4\omega / (16 - \omega^2)]]$$

Q3

(15 marks)

Consider a control system shown in Fig.3 if

$$G(s)H(s) = \frac{k}{(s+1+j)(s+2)(s+1-j)} = \frac{k}{s^3 + 4s^2 + 6s + 4}$$

- Sketch the **complete root locus** for positive values of **K**?
- Find **K** that makes the complex closed loop poles have a damping ratio =**0.5** and **find the closed loop poles** using **the plot**?
- Find **K** that makes the complex closed loop poles have a damping ratio =**0.5** and **find the closed loop poles analytically**?
- Write short MATLAB program to solve a & b?

a-Root locus:

1-the root locus is symmetrical about the real axis in the S-plane

$$2\text{-the open loop } G(s)H(s) = \frac{k}{(s+1-j)(s+2)(s+1+j)} = \frac{k}{s^3 + 4s^2 + 6s + 4}$$

3-the root locus starts at the pole and ends at the zeros or infinity

4-number of root loci= n=number of poles of the open loop TF =3 at  $[-1+j, -1-j, -2]$

5-number of zeros= m=0

6-number of asymptotes = n-m=3-0=3

8-center of gravity =  $A = \frac{\sum \text{poles} - \sum \text{zeros}}{n-m} = \frac{-1-1-2}{3} = -1.3$  point of intersection of asymptotes with real axis=

9-angles of asymptotes are =  $\theta = \frac{\pm 180(2R+1)}{n-m} = \pm 60, \pm 180$

10- Points of crossing the imaginary axis as Routh test

Charct. equa=1+G(S)H(S)=0=  $S^3+4S^2+6s+4+K$

$S^3$	1	6	$4+K \geq 0, [20-K]/4 \geq 0$ then $-4 \leq K \leq 20, K_c = 20$ $4S^2+24=0, S = \pm j\omega = \pm \sqrt{6}$ rad/sec
$S^2$	4	4+K	
S	$[24-4-K]/4$		
$S^0$	4+K		

11- there is no break points (break away or break in) at

$$-\frac{dK}{dS} = 0 = \frac{d}{dS} \left[ \frac{1}{G(S)H(S)} \right] = \frac{d}{dS} [S^3 + 4S^2 + 6s + 4] = 3S^2 + 8s + 6 = 0$$

12- There is no break angles  $[\pm 180(2R+1)/r]$  where r=number of branches (poles for break away or zeros for break in) R=0,1,-----no break angles

13-the angle of departures (complex poles) =

**Angle of departure from a complex pole – 180°**

– (sum of the angles of vectors to a complex pole in question from other poles)

+ (sum of the angles of vectors to a complex pole in question from zeros)

angle of departure=  $\pm 180 - 90 - 45 = \pm 45$  deg

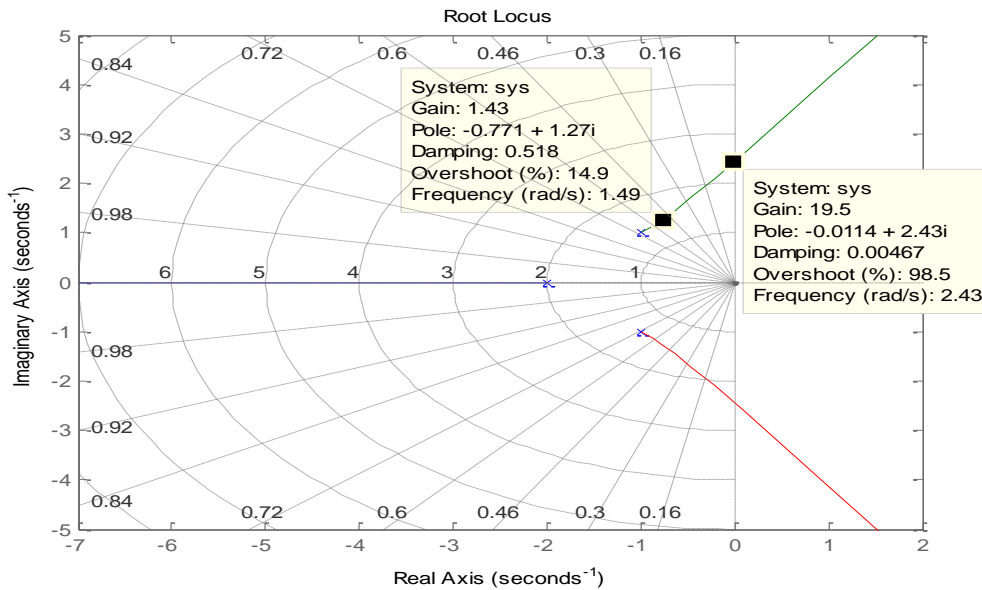
14-angle of arrival (complex zeros) as

**Angle of arrival at a complex zero = 180°**

– (sum of the angles of vectors to a complex zero in question from other zeros)

+ (sum of the angles of vectors to a complex zero in question from poles)

15-sketch the root loci as



16- the damping factor or coefficient  $\zeta$  is straight line with slope  $\Theta = \cos^{-1}\zeta$

with respect to the negative real axis in the S-plane.  $\Theta = \cos^{-1} 0.5 = 60\text{deg}$ . at the test point (intersection point)  $S_d = -0.8 \pm j1.3$

$$\text{angle condition} = \sum_{n=1}^{n=3} [\theta_{zeros} - \theta_{poles}] = \pm 180(2R + 1) = 90 + 54 + 36 = 180 \text{ deg}$$

$$\text{magnitude condition} = \sum_{n=1}^{n=3} \frac{\|poles\|}{\|zeros\|} = K = * * = 1.5$$

$$\sum_{n=1}^{n=3} \text{open loop poles} = \sum_{n=1}^{n=3} \text{closed loop poles} = \text{constant as } n - m \geq 2$$

$$\sum_{n=1}^{n=3} \text{open loop poles} = -1 - 2 - 1 = -4 = \sum_{n=1}^{n=3} \text{closed loop poles} \\ = (-0.8 + j1.3, -0.8 - j1.3, p)$$

then  $p = -2.4$  i. e. closed loop poles are  $[-0.8 \pm j1.3, -2.4]$

**19- To find analytically closed loop poles and K as**

$(S^2 + 2\zeta\omega_n S + \omega_n^2)(S+a) = \text{characteristic equa. for a third order syst.}$

Solve  $1 + G(S)H(S) = 0 = S^3 + 4S^2 + 6s + 4 + K = (S^2 + \omega_n S + \omega_n^2)(S+a)$



$$= S^3 + (\omega_n + a)S^2 + (\omega_n a + \omega_n^2)S + \omega_n^2 a$$

$$\omega_n + a = 4, \omega_n a + \omega_n^2 = 6, \omega_n^2 a = k + 4, \text{ then } \omega_n = 1.5 \text{ rad/sec}, a = 2.5, k = 1.74$$

Prog.  $\gg n=[1]; d=[1 \ 4 \ 6 \ 4]; \text{ rlocus}(n,d), \text{ grid}$

Q4

(30 marks)

a- Define: gain margin- phase margin?

**-Gain margin  $G_m$ :** it is reciprocal of the magnitude of the output frequency response at the **Phase crossover frequency  $\omega_p$**

$$G_m = 1/[\text{Real of } G(j \omega_p)H(j \omega_p)] = 1/|G(j \omega_p)H(j \omega_p)| = K_c/K$$

$$G_M = 20 \log G_m \text{ db}$$

**-Phase margin  $\gamma_m$ :** it is the angle of the output frequency response at the **gain crossover** frequency plus 180 degrees.

$$\gamma_m = \angle G(j \omega_g)H(j \omega_g) + 180 \text{ deg.}$$

b- Define:  $\omega_n, \omega_d, \omega_r, \omega_c, \omega_g, \omega_p, M_r, \eta$ ?

**-Natural frequency  $\omega_n$  rad/sec:** it is the natural frequency depends on the natural of the system parameters.

**- Under damped natural frequency  $\omega_d$  rad/sec:** it is the under damped natural frequency depends on the damping coefficient  $\eta$  as it is less than one  $\eta < 1$ .

**-Resonant frequency  $\omega_r$  rad/sec:** it is the frequency at which the peak value of the output frequency response for a second order is equal to  $\omega_r = \omega_n \sqrt{1 - 2\zeta^2}$

$$\omega_r = \omega_n \sqrt{1 - 2\zeta^2}, \quad \text{for } 0 \leq \zeta \leq 0.707$$

**As  $\zeta$  approaches zero,  $M_r$  approaches infinity**

$0 < \zeta \leq 0.707$ , the resonant frequency  $\omega_r$  is less than the damped natural frequency

**-Corner frequency  $\omega_c$  rad/sec:** it is the frequency at which the magnitude of the output frequency response is changed sharply. It may be  $(0, 1, 1/T, \omega_n)$

**-Gain crossover frequency  $\omega_g$ :** it is the frequency at which the magnitude of the output frequency response is equal to one or zero decibel.

$$|G(j \omega_g)H(j \omega_g)| = 1 \quad \text{or} \quad |G(j \omega_g)H(j \omega_g)| = 0 \text{ db}$$

**-Phase crossover frequency  $\omega_p$ :** it is the frequency at which the phase of the output frequency response is equal to  $(-180)$  degrees.

$$\text{Imag. } [G(j \omega_p)H(j \omega_p)] = 0 \quad \text{or} \quad \angle G(j \omega_p)H(j \omega_p) = -180 \text{ deg.}$$

**-Maximum resonant magnitude  $M_r$ :** it is the peak value of the output frequency response for a second order system  $M_r = \frac{1}{2\zeta\sqrt{1-\zeta^2}}$

$$\frac{C(s)}{R(s)} = \frac{\omega_n^2}{s^2 + 2\zeta\omega_n s + \omega_n^2} \quad M_r = |G(j\omega)|_{\max} = |G(j\omega_r)| = \frac{1}{2\zeta\sqrt{1-\zeta^2}}$$

**-damping coefficient  $\eta$**  it depends on the natural of the system parameters. For second order system

$$\frac{C(s)}{R(s)} = \frac{\omega_n^2}{s^2 + 2\zeta\omega_n s + \omega_n^2}$$

Values of $\eta$	System stability	Step-response
$0 > \eta$	System is unstable	undefined
$\eta = 0$	System is critically stable	oscillatory
$0 < \eta < 1$	System is stable	Under-damped
$0 < \eta = 1$	System is stable	Critically damped
$0 < \eta > 1$	System is stable	Over damped

c-Consider a control system shown in Fig.1 if

$$G(s)H(s) = \frac{10}{(S + 1 + j)(S + 2)(S + 1 - j)} = \frac{10}{S^3 + 4S^2 + 6S + 4}$$

- Prove that the gain margin=**6.02 db at 2.45 rad/sec.** and the phase margin=**30.3 degrees at 1.78 rad/sec.?**
  - Sketch the **polar plot?**
  - Sketch the **Bode plot?**
  - Show the gain margin and the phase margin on **the plots?**
  - Write short MATLAB program to solve a , b and C?
- 1- the open loop TF= $G(s) H(s)= \mathbf{G(S) H(S)}$

$$G(s)H(s) = \frac{10}{(S + 1 + j)(S + 2)(S + 1 - j)} = \frac{10}{S^3 + 4S^2 + 6S + 4}$$

- 2- Find the freq.open loop TF=

$$\mathbf{G(j\omega)H(j\omega)} = \frac{10}{S^3 + 4S^2 + 6S + 4} = M e^{j\Phi} = M \angle \Phi = \text{Re} + j \text{imag}$$

$$M = \frac{10}{\sqrt{(4-4\omega^2)^2+(6\omega-\omega^3)^2}}, \Phi = -\tan^{-1}((6\omega - \omega^3)/(4 - 4\omega^2))$$

$$M = \frac{10}{\sqrt{(4-4\omega^2)^2+(6\omega-\omega^3)^2}} = \frac{10}{\sqrt{(4-4\omega^2)^2+(6\omega-\omega^3)^2}} = 1$$

$$M = \frac{10}{\sqrt{(4-4(1.78)^2)^2+(6(1.78)-(1.78)^3)^2}} = 1, \text{ then } \omega_g = 1.78\text{rad/sec.}$$

$$\Phi = -\tan^{-1}((6\omega - \omega^3)/(4 - 4\omega^2)) = -\tan^{-1}((6 * 2.45 - 2.45^3)/(4 - 4 * 2.45^2)) = -180 \text{ deg.}$$

then  $\omega_p = 2.45\text{rad/sec.}$

$$M = \frac{10}{\sqrt{(4-4(2.45)^2)^2+(6(2.45)-(2.45)^3)^2}} = 0.5, \text{ then } G_M = 20\log \frac{1}{0.5} = 6.02\text{db}$$

$$\Phi = -\tan^{-1}((6\omega - \omega^3)/(4 - 4\omega^2)) = -\tan^{-1}((6 * 1.78 - 1.78^3)/(4 - 4 * 1.78^2)) = -149.7\text{deg.}$$

$$\gamma_m = \angle G(j\omega_g)H(j\omega_g) + 180 \text{ deg.} = 180 - 149.7 = 30.3\text{deg.}$$

3- Find the table

$\omega$	0	0.1	1	1.78	2.45	5	10	$\infty$
$\Phi$	0			-150	-180	-		-270
M	2.5		2	1	0.5			0
20logM	8		6.02-	0	6.02	-	-	0
Real $G(j\omega)H(j\omega)$	2.5		0		-0.5			0
Imag $G(j\omega)H(j\omega)$	0		-2		0			0

4- Plot the vector on the  $j\omega$  - plane where  $\Phi$  in degrees as a straight line and determine M on this line

5- Plot the locus of the vector as points from the table

6- Find the gain and the phase margins from the plot

**Prog.** `>>n=[10]; d=[1 4 6 4];`

`>> nyquist(n,d)    >> margin(n,d)    >> nichols(n,d)`

