



Model Answer of the Final Exam

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المادة : التحكم الآلي م 482

أستاذ المادة : د. محمد عبد اللطيف الشرنوبى

1-a) Find the signal flow graph of the system shown in Figure 1 Use the given variables as nodes. List all loops.

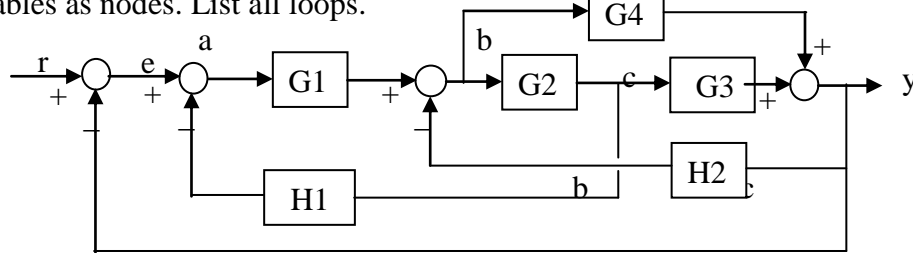
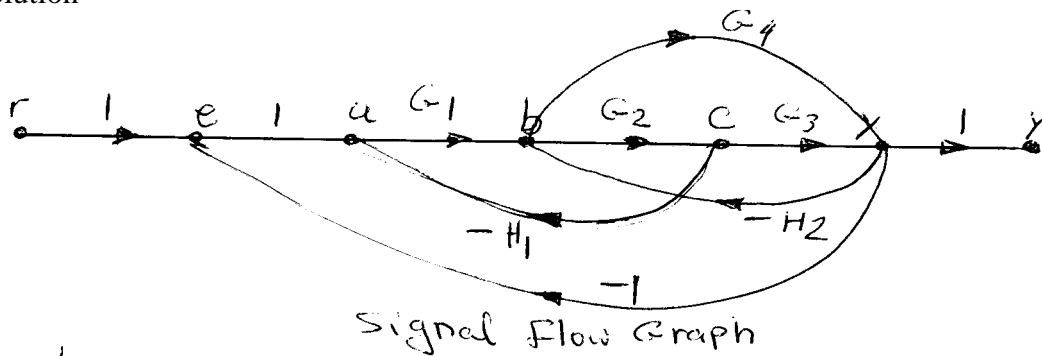


Figure 1

Solution



Loops

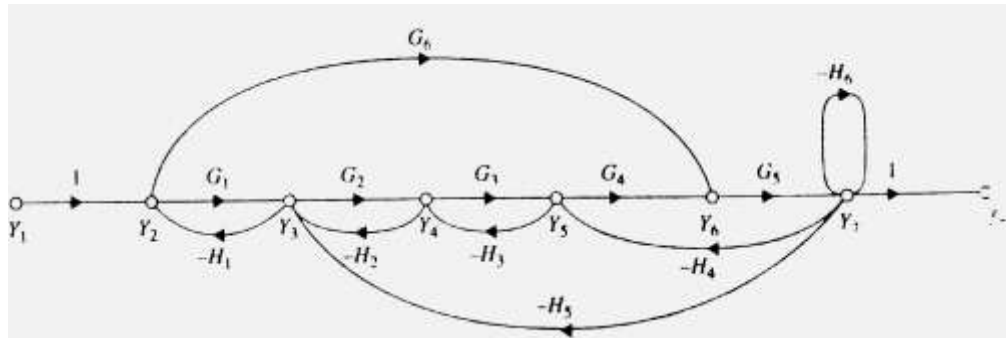
$$L_1 = -G_1 G_2 H_1 \quad L_2 = -G_2 G_3 H_2$$

$$L_3 = -G_4 H_2 \quad L_4 = -G_1 G_2 G_3$$

$$L_5 = -G_1 G_4$$

1-b)

The SFG is given by:



Solution

Loops

$$L_1 = -G_1 H_1 \quad L_2 = -G_2 H_2 \quad L_3 = -G_3 H_3$$

$$L_4 = -G_4 G_5 H_4 \quad L_5 = -H_6 \quad L_6 = -G_2 G_3 G_4 G_5 H_5$$

Paths

$$P_1 = G_1 G_2 G_3 G_4 G_5 \quad P_2 = G_6 G_5$$

$$\Delta = 1 - \sum_{i=1}^6 L_i + \sum_{\substack{i,j \\ \text{non touching}}} L_i L_j - \sum_{\substack{i,j,k \\ \text{non touching}}} L_i L_j L_k$$

$$= 1 + G_1 H_1 + G_2 H_2 + G_3 H_3 + G_4 G_5 H_4 + H_6 + G_2 G_3 G_4 G_5 H_5$$

$$+ L_1 L_3 + L_1 L_4 + L_1 L_5 + L_2 L_4 + L_2 L_5 + L_3 L_5 + L_3 L_6$$

$$- L_1 L_3 L_5$$

$$\Delta = 1 + (G_1 H_1 + G_2 H_2 + G_3 H_3 + G_4 G_5 H_4 + H_6 + G_2 G_3 G_4 G_5 H_5)$$

$$+ G_1 H_1 G_3 H_3 + G_1 H_1 G_4 G_5 H_4 + G_1 H_1 H_6 + G_2 H_2 G_4 G_5 H_4$$

$$+ G_2 H_2 H_6 + G_3 H_3 H_6 + G_2 G_3^2 G_4 G_5 H_3 H_5$$

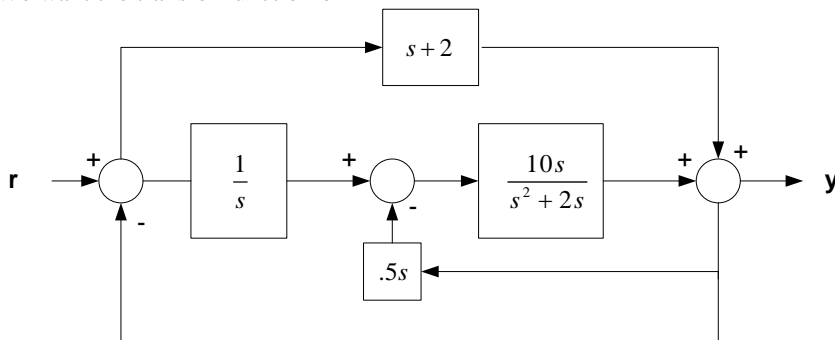
$$+ G_1 H_1 G_3 H_3 H_6$$

$$\Delta_1 = 1 \quad \Delta_2 = 1 + G_2 H_2 + G_3 H_3$$

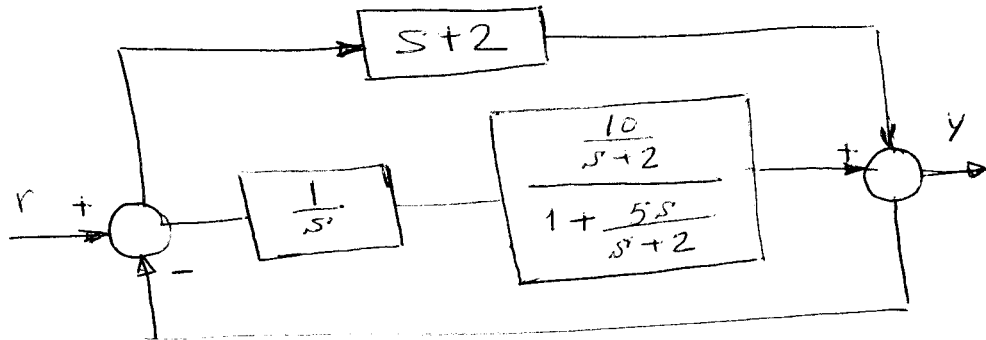
$$TF = \frac{P_1 \Delta_1 + P_2 \Delta_2}{\Delta} = \frac{G_1 G_2 G_3 G_4 G_5 + G_5 G_6 (\Delta_2)}{\Delta}$$

2-a)

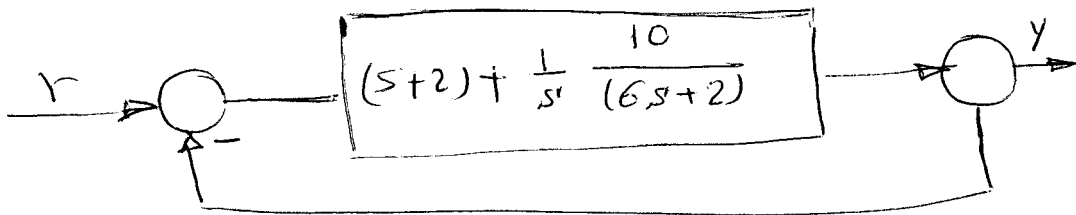
We want the transfer function of



The Block $\frac{10s}{s^2+2s} = \frac{10}{s+2}$



The system is reduced to



(i) T.F =
$$\frac{(s+2) + \frac{5}{3s^2+s}}{1 + (s+2) + \frac{5}{3s^2+s}}$$

$$= \frac{3s^3 + 7s^2 + 2s + 5}{3s^3 + 10s^2 + 3s + 5}$$

If r is unit step, The open loop T.F

$$G(s) = (s+2) + \frac{1}{s} \left(\frac{10}{6s+2} \right)$$

(ii) $K_p = \lim_{s \rightarrow 0} G(s) = \infty \Rightarrow e_{ss} = \frac{1}{1+K_p} = 0$

(iii) $K_v = \lim_{s \rightarrow 0} sG(s) = 5 \Rightarrow e_{ss} = \frac{1}{5} = 20\%$

Construct The Howartz array

| | | |
|-------|-----|---|
| s^3 | 3 | 3 |
| s^2 | 10 | 5 |
| s | 1.5 | 0 |
| 1 | 5 | 0 |

The system is

BIBO stable.

2-b) Hand sketch the root locus of $1 + KG(s) = 0$ as K varies from 0 to $+\infty$, where

$$G(s) = \frac{s+2}{s(s+1)(s+3)^2}$$

open loop poles are 0, -1, -3, -3

" " zero are -2 =

Number of asymptotes are $n - m = 3$

$$\text{asymptotes angle} = \frac{(2k+1)\pi}{3} = 60^\circ, 180^\circ, 300^\circ$$

Intersection of Asymptotes on the real line

$$\sigma = \frac{\sum \text{Poles} - \sum \text{Zeros}}{n - m} = \frac{-1 - 3 - 3 + 2}{3} = \frac{-5}{3}$$

The characteristic eqn. is

$$1 + KG(s) = 0 \Rightarrow 1 + \frac{k(s+2)}{(s^2+s)(s^2+6s+9)} = 0$$

$$s^4 + 7s^3 + 15s^2 + 9s + ks + 2k = 0$$

Construct the array.

$$\begin{array}{r|rrrr} s^4 & 1 & 15 & 2k & \\ s^3 & 7 & (9+k) & 0 & \\ s^2 & \frac{105-9k-9}{7} & 2k & & \\ s & \frac{(96-9k)(9+k)}{96-9k} & -98k & 0 & \\ 1 & & 2k & & \end{array}$$

$$96 - 9k > 0 \Rightarrow k < \frac{96}{9}$$

$$(96 - 9k)(9+k) - 98k > 0 \Rightarrow$$

$$864 - 81k - 98k - 9k^2 > 0$$

$$864 - 179k - 9k^2 > 0 \quad k \leq 4,$$

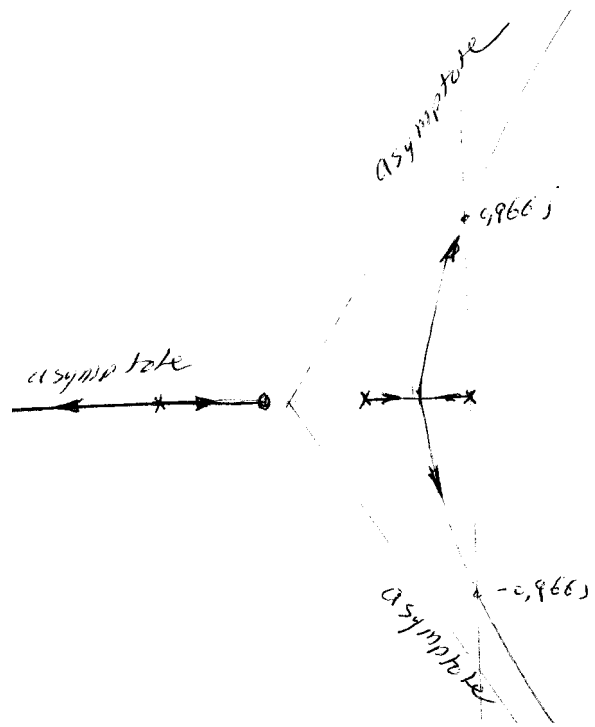
For $k = 4$

The Auxiliary Eqn. is $\frac{60}{7}s^2 + 8 = 0$

$$s = \pm \sqrt{\frac{14}{15}} j$$

$$= \pm 0.966j$$

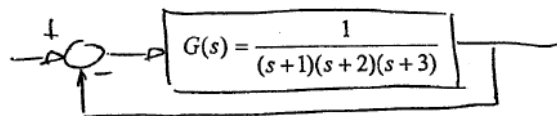
points intersect
The Imaginary
line.



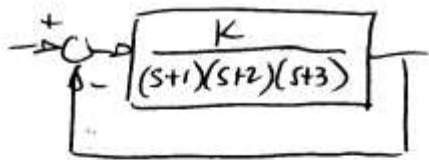
The root locus

3-a)

Given the control system



- (i) Analytically determine, as shown in class, the 180 degree phase crossover point of the Nyquist diagram and the corresponding frequency at the 180 degree phase crossover point.
- (ii) Assuming that the 180 degree phase crossover point is -0.5 , what would the gain margin of the system be?



$$1 + \frac{K}{(s+1)(s+2)(s+3)} = 0$$

$$\Rightarrow 1s^3 + 6s^2 + 11s + (6+K)$$

(a)

| | | |
|-------|------|-----|
| s^3 | 1 | 11 |
| s^2 | 6 | 6+K |
| s^1 | 60-K | |
| s^0 | 6+K | |

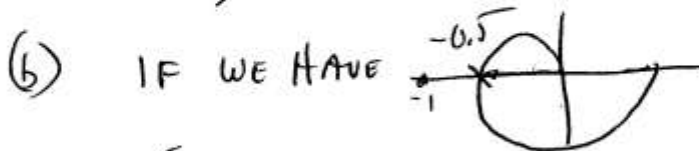
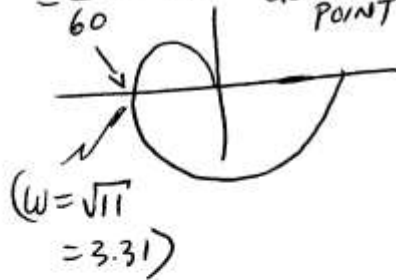
$$60 - K = 0 \Rightarrow K = 60$$

$$\text{Aux. Eq. : } 6s^2 + 6 + 60 = 0$$

$$s^2 + 11 = 0 \Rightarrow$$

$$s = \pm j\sqrt{11}$$

$-\frac{1}{60} = \text{THE } 180^\circ \text{ PHASE CROSSOVER POINT.} = -0.0167$



$$(G.M.)(-0.5) = -1 \Rightarrow \underline{\underline{GM = 2}}$$

3-b)
4-a)

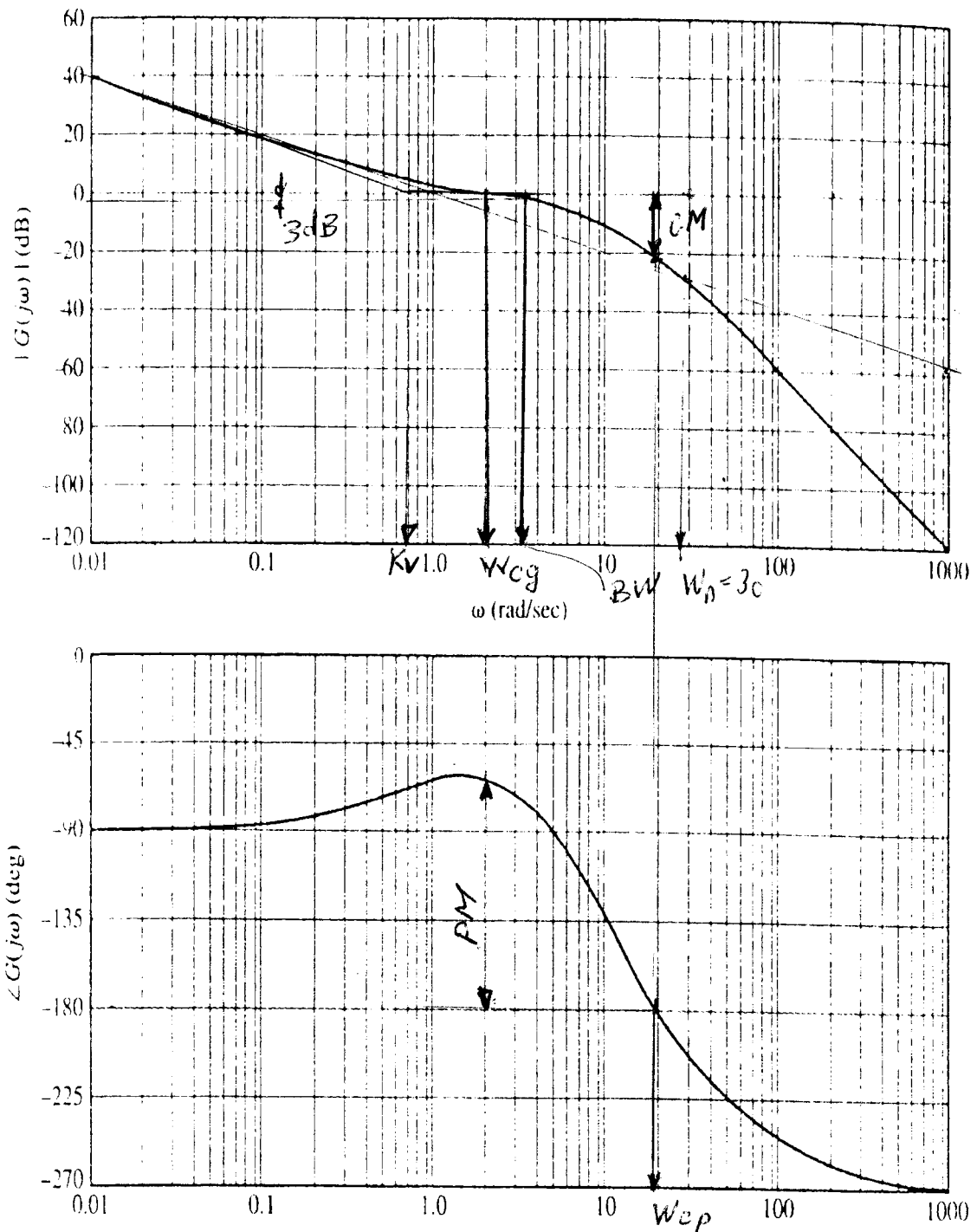


Figure 4

$$\begin{aligned} \omega_{cg} &= 2 \text{ rad/sec} \\ \omega_{cp} &= 19.5 \text{ rad/sec} \\ GM &= 22 \text{ dB} \\ PM &= 115^\circ \\ \omega_n &= 30 \text{ rad/sec} \end{aligned}$$

$$\begin{aligned} K_V &= 0.9 \sim 1.0 \\ \text{system type } &1 \\ e_{ss} &= \frac{1}{K_V} \approx 1 \sim 1.11 \\ BW &= 2.4 \text{ rad/sec} \\ BW &= 2.4 \text{ rad/sec} \end{aligned}$$

ii) The closed loop system is stable when applying Nyquist Criteria. As GM is positive, the polar plot does not contain the critical point. All required values are above

4-a) 3-b) For the characteristic equation of the feedback system given by:

$$s^4 + Ks^3 + 5s^2 + 10s + 10K = 0$$

Determine the range of k so that the system is stable.

Construct Routh array.

| | | | |
|-------|----------------------------|-------|-------|
| s^4 | 1 | 5 | $10k$ |
| s^3 | k | 10 | 0 |
| s^2 | $\frac{5k-10}{k}$ | $10k$ | |
| s^1 | $\frac{10 - 10k^3}{5k-10}$ | 0 | |
| 1 | $10k$ | | |

① For stable system
 $5k - 10 > 0 \quad k > 2$

② $10 - \frac{2k^3}{k-2} > 0$

Condition 2 will not be satisfied for any $k \rightarrow$ The system is unstable for any k

4-b) For the system given below in figure 5, estimate the values of K and K_t so that a maximum percentage overshoot of 9.6% and a settling time of 0.05 sec for a tolerance band of 1% are achieved.

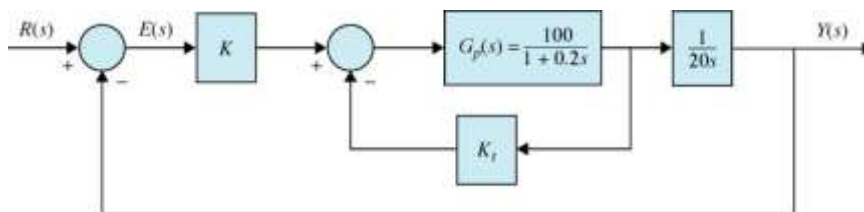
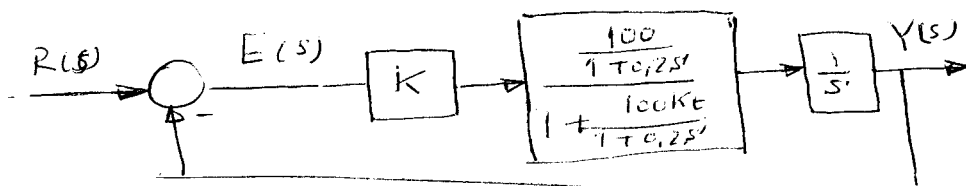
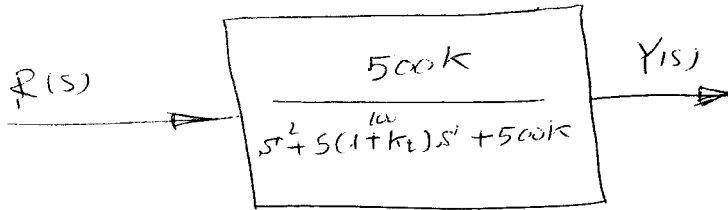
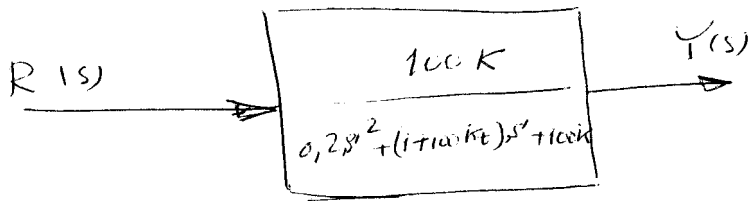
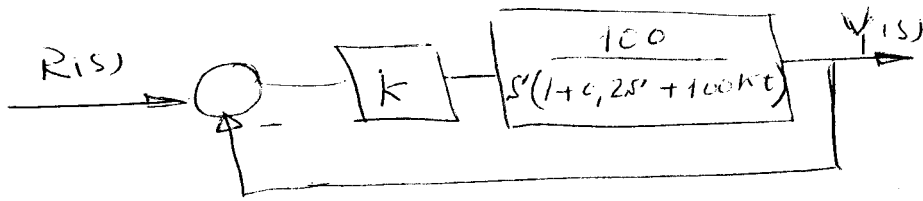


Figure 5

The system is reduced to





$$\omega_n^2 = 500k$$

$$2\zeta\omega_n = (5 + 500k_t)$$

for 1% settling time we know

$$t_{st} = \frac{4.6}{\zeta\omega_n} = 0,05$$

$$\therefore \zeta\omega_n = 92$$

$$\therefore k_t = 0,358$$

$$M_p = \frac{1}{2\zeta\sqrt{1-\zeta^2}} = 0,096$$

$$= e^{-\pi\zeta/\sqrt{1-\zeta^2}} = 0,096$$

$$\zeta = 0,5979 \Rightarrow \omega_n = 153,87$$

$$k = \frac{\omega_n^2}{500} = 47,35$$

5) Consider the standard feedback system shown below in figure 6.

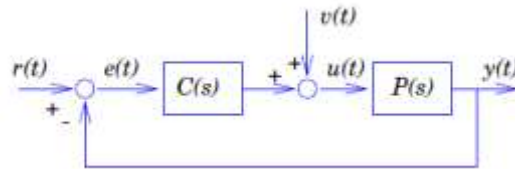
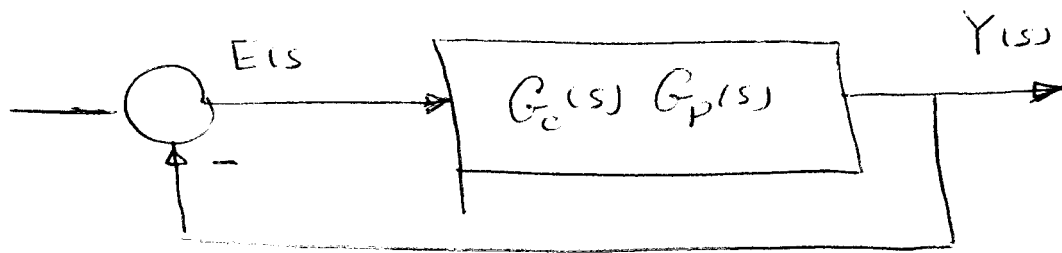


Figure 6



The characteristic Eqn. is given by:

$$1 + G_c(s) G_p(s) = 0$$

$$1 + \frac{k_p s + k_i}{s} \frac{1}{(s^2 + 5s)} = 0$$

$$s^3 + 5s^2 + k_p s + k_i = 0$$

$s = -4$ satisfies the eqn.

$$-64 + 80 - 4k_p + k_i = 0$$

$$16 - 4k_p + k_i = 0$$

$$k_i = 4k_p - 16$$

$$s^3 + 5s^2 + k_p s + 4k_p - 16 = 0$$

$$k_p > 4$$

$$(s^3 + 4s^2) + (s^2 - 16) + K_P(s+4) = 0$$

$$s^2(s+4) + (s+4)(s-4) + K_P(s+4) = 0$$

$$(s+4) [s^2 + s + (K_P - 4)] = 0$$

$$r_{2,3} = \frac{-1 \pm \sqrt{1+16-4K_P}}{2}$$

$$= -\frac{1}{2} \pm j \sqrt{K_P - \frac{17}{4}} = -\zeta \omega_n \pm j \omega_n \sqrt{1-\zeta^2}$$

choose $\omega_n = 2 \rightarrow \zeta = 0,25$

$$4(1-\zeta^2) = K_P - \frac{17}{4} \Rightarrow \frac{15}{4} + \frac{17}{4} = K_P \Rightarrow$$

$$\boxed{K_P = 8}$$

The transfer function $\frac{1}{s^2(s+5)}$

$$= \frac{1}{5} \frac{1}{(s)} \frac{1}{(0,25s+1)}$$